

#### **Features**

- Constant-Current/Constant-Voltage Operation with Thermal Regulation to Maximize Charge Rate Without Risk of Overheating
- No MOSFET, Sense Resistor or Blocking Diode Required
- Complete Linear Charger in SOT Package for Single Cell Lithium-Ion Batteries
- Charges Single Cell Li-Ion Batteries Directly from USB Port
- Charge Current Monitor Output for Gas Gauging

### Applications

- Cellular Telephones
- Charging Docks and Cradles

## **General Description**

- Automatic Recharge
- Charge Status Output Pin
- C/10 Charge Termination
- Programmable Charge Current Up to 500mA
- Preset 4.2V Charge Voltage with ±1% Accuracy
- 25µA Supply Current in Shutdown
- 2.9V Trickle Charge Threshold
- Soft-Start Limits Inrush Current
- Available in 5-Lead SOT-23 Package
- RoHS Compliant and Lead (Pb) Free
- Bluetooth Application
- Wearable Application

The RY4054 is a complete constant-current/constant-voltage linear charger for single cell lithium-ion batteries. Its SOT package and low external component count make the RY4054 ideally suited for portable applications. Furthermore, the RY4054 is specifically designed to work within USB power specifications.

No external sense resistor is needed, and no blocking diode is required due to the internal MOSFET architecture. Thermal feedback regulates the charge current to limit the die temperature during high power operation or high ambient temperature. The charge voltage is fixed at 4.2V, and the charge current can be programmed externally with a single resistor. The RY4054 automatically terminates the charge cycle when the charge current drops to 1/10th the programmed value after the final float voltage is reached.

When the input supply (wall adapter or USB supply) is removed, the RY4054 automatically enters a low current state, dropping the battery drain current to less than  $2\mu A$ . The RY4054 can be put into shutdown mode, reducing the supply current to  $25\mu A$ . Other features include charge current monitor, undervoltage lockout, automatic recharge and a status pin to indicate charge termination and the presence of an input voltage.

## **Typical Application Circuit**

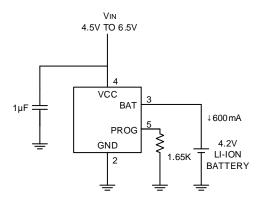


Figure 1. Typical Application Circuit



## **Pin Description**

#### **Pin Configuration**

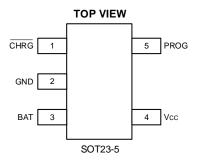


Figure 2. Pin Configuration

Top Marking: NAYLL (device code: NA, Y=year code, LL= lot number code)

#### **Pin Description**

Pin	Name	Function
1	CHRG	Open-Drain Charge Status Output Pin
2	GND	Ground
3	BAT	Charge Current Output Pin
4	VCC	Positive Input Supply Voltage Pin
5	PROG	Charge Current Program, Charge Current Monitor and Shutdown Pin

## **Order Information**

Marking	Part No.	Model	Description	Package	T/R Qty
NA <u>YLL</u>	70705001	RY4054AT5	RY4054AT5 Linear Li-Ion Battery Charger, 4.3-6.5V, 500mA, Preset 4.2V, IQ 25µA, 1 Charge Status, SOT23-5	SOT23-5	3000 PCS

Note (1): All RYCHIP parts are Pb-Free and adhere to the RoHS directive.



## **Specifications**

#### Absolute Maximum Ratings (1) (2)

Item	Min	Max	Unit
VCC Input Voltage	-0.3	10	V
PROG Voltage	-0.3	VCC+0.3	V
BAT Voltage	-0.3	7	V
CHRG Voltage	-0.3	10	V
Power dissipation	Internally Lim	ited	
Operating Junction Temperature, T <sub>J</sub>	-40	150	°C
Storage Temperature, T <sub>stg</sub>	-65	125	°C
Operating Temperature Range	-40	85	°C
Lead Temperature (Soldering, 10sec.)		260	°C

Note (1): Exceeding these ratings may damage the device.

Note (2): The device is not guaranteed to function outside of its operating conditions.

#### **ESD Ratings**

Item	Description	Value	Unit
	Human Body Model (HBM)		
V <sub>(ESD-HBM)</sub>	ANSI/ESDA/JEDEC JS-001-2014	$\pm 2000$	V
	Classification, Class: 2		
	Charged Device Model (CDM)		
V <sub>(ESD-CDM)</sub>	ANSI/ESDA/JEDEC JS-002-2014	±200	V
	Classification, Class: C0b		
	JEDEC STANDARD NO.78E APRIL 2016		
I <sub>LATCH-UP</sub>	Temperature Classification,	±150	mA
	Class: I		

#### **Recommended Operating Conditions**

Item	Min	Max	Unit
Operating junction temperature <sup>(1)</sup>	-40	125	°C
Operating temperature range	-40	85	°C

Note (1): All limits specified at room temperature ( $TA = 25^{\circ}C$ ) unless otherwise specified. All room temperature limits are 100% production tested. All limits at temperature extremes are ensured through correlation using standard Statistical Quality Control (SQC) methods. All limits are used to calculate Average Outgoing Quality Level (AOQL).

#### **Thermal Information**

Item	Description	Value	Unit
$R_{\theta JA}$	Junction-to-ambient thermal resistance <sup>(1)(2)</sup>	180	°C/W
R <sub>0</sub> JC(top)	Junction-to-case (top) thermal resistance	130	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	45	°C/W
ΨJT	Junction-to-top characterization parameter	35	°C/W



ψ <sub>JB</sub> Junction-to-board characterization parameter 45	°C/W
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Note (1): The package thermal impedance is calculated in accordance to JESD 51-7.

Note (2): Thermal Resistances were simulated on a 4-layer, JEDEC board.

## Electrical Characteristics <sup>(1) (2) (3)</sup>

Vaa-	5	ТΛ —	25°C	unlage	otherwise noted.
$v_{\rm CC} -$	э,	IA -	23 C,	umess	otherwise noted.

Parameter	Symbol	Test Conditions	Min	Тур.	Max	Unit
Input Supply Voltage	V <sub>CC</sub>		4.25		6.5	V
		Charge Mode (Note 4), $R_{PROG} = 10k$		300	2000	μΑ
Input Supply Current	т	Standby Mode (Charge Terminated)		200	500	μΑ
input Suppry Current	I <sub>CC</sub>			25	50	μΑ
Regulated Output (Float) Voltage	V <sub>FLOAT</sub>	$0^{\circ}C \leq TA \leq 85^{\circ}C, I_{BAT} = 40mA$	4.158	4.2	4.242	V
		$R_{PROG} = 10k$ , Current Mode	93	100	107	mA
		$R_{PROG} = 2k$ , Current Mode	465	500	535	mA
BAT Pin Current	I	Standby Mode, $V_{BAT} = 4.2V$	0	-2.5	-6	μΑ
BAI Pin Current	I <sub>BAT</sub>	Shutdown Mode (R <sub>PROG</sub> Not Connected)		±1	±2	μΑ
		Sleep Mode, $V_{CC} = 0V$		±1	±2	μΑ
Trickle Charge Current	I <sub>trikl</sub>	$V_{BAT} < V_{TRIKL}, R_{PROG} = 2k$	20	45	70	mA
Trickle Charge Threshold Voltage	V <sub>TRIKL</sub>	R <sub>PROG</sub> = 10k, VBAT Rising	2.8	2.9	3.0	V
Trickle Charge Hysteresis Voltage	V <sub>TRHYS</sub>	$R_{PROG} = 10k$	60	80	110	mV
V <sub>CC</sub> Undervoltage Lockout Threshold	$V_{\rm UV}$	From $V_{CC}$ Low to High	3.7	3.8	3.92	V
V <sub>CC</sub> Undervoltage Lockout Hysteresis	$V_{\rm UVHYS}$		150	200	300	mV
Manual Shutdown	V <sub>MSD</sub>	PROG Pin Rising	1.15	1.21	1.30	V
Threshold Voltage	V MSD	PROG Pin Falling	0.9	1.0	1.1	V
$V_{CC} - V_{BAT}$ Lockout	V <sub>ASD</sub>	V <sub>CC</sub> from Low to High	70	100	140	mV
Threshold Voltage	v ASD	V <sub>CC</sub> from High to Low	5	30	50	mV
C/10 Termination	I <sub>term</sub>	$R_{PROG} = 10k \text{ (Note 5)}$	0.085	0.10	0.115	mA/ mA
Current Threshold		$R_{PROG} = 2k$	0.085	0.10	0.115	mA/ mA
PROG Pin Voltage	VPROG	R <sub>PROG</sub> = 10k, Current Mode	0.93	1.0	1.07	V
<i>CHRG</i> Pin Weak Pull-Down Current	I <sub>CHRG</sub>	$V_{CHRG} = 5V$	8	20	35	μΑ
CHRG Pin Output	V <sub>CHRG</sub>	$I_{CHRG} = 5mA$		0.35	0.6	V

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Low Voltage						
RechargeBatteryThresholdVoltageProtection	$\Delta V_{RECHRG}$	Vfloat - Vrechrg	100	150	200	mV
Junction Temperature						
in Constant Temperature Mode	$T_{\text{LIM}}$			120		°C
Power FET "ON"						
Resistance (Between	R <sub>ON</sub>			600		mΩ
V <sub>CC</sub> and BAT)						
Soft-Start Time	$T_{SS}$	$I_{BAT}\!=\!0$ to $I_{BAT}\!=\!1000V\!/\!R_{PROG}$		100		μS
Recharge Comparator Filter Time	<i>t</i> recharge	V <sub>BAT</sub> High to Low	0.75	2	4.5	mS
Termination						
Comparator Filter	t <sub>TERM</sub>	I <sub>BAT</sub> Falling Below I <sub>CHG</sub> /10	400	1000	2500	μS
Time						
PROG Pin Pull-Up Current	I <sub>PROG</sub>			3		μΑ

Note (1): Absolute Maximum Ratings are those values beyond which the life of the device may be impaired. Note (2): The RY4054 is guaranteed to meet performance specifications from 0°C to 70°C. Specifications over the -40°C to 85°C operating temperature range are assured by design, characterization and correlation with statistical process controls.

Note (3): See Thermal Considerations.

Note (4): Supply current includes PROG pin current (approximately  $100\mu A$ ) but does not include any current delivered to the battery through the BAT pin (approximately 100mA).

Note (5): I<sub>TERM</sub> is expressed as a fraction of measured full charge current with indicated PROG resistor.





### **Functional Block Diagram**

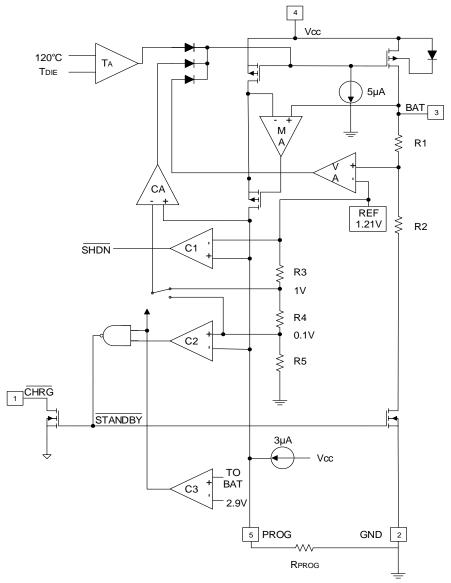


Figure 3. Functional Block Diagram

### **Functions Description**

The RY4054 is a single cell lithium-ion battery charger using a constant-current/constant-voltage algorithm. It can deliver up to 500mA of charge current (using a good thermal PCB layout) with a final float voltage accuracy of  $\pm$ 1%. The RY4054 includes an internal P-channel power MOSFET and thermal regulation circuitry. No blocking diode or external current sense resistor is required; thus, the basic charger circuit requires only two external components. Furthermore, the RY4054 is capable of operating from a USB power source.

#### Normal Charge Cycle

A charge cycle begins when the voltage at the VCC pin rises above the UVLO threshold level and a 1% program resistor is connected from the PROG pin to ground or when a battery is connected to the charger output. If the BAT pin is less than 2.9V, the charger enters trickle charge mode. In this mode, the RY4054 supplies approximately 1/10 the programmed charge current to bring the battery voltage up to a safe level for full current charging. (Note: The



RY4054X does not include this trickle charge feature).

When the BAT pin voltage rises above 2.9V, the charger enters constant-current mode, where the programmed charge current is supplied to the battery. When the BAT pin approaches the final float voltage (4.2V), the RY4054 enters constant-voltage mode and the charge current begins to decrease. When the charge current drops to 1/10 of the programmed value, the charge cycle ends.

#### **Programming Charge Current**

The charge current is programmed using a single resistor from the PROG pin to ground. The battery charge current is 1000 times the current out of the PROG pin. The program resistor and the charge current are calculated using the following equations:

$$R_{PROG} = \frac{1000V}{I_{CHG}}$$
$$I_{CHG} = \frac{1000V}{R_{PROG}}$$

The charge current out of the BAT pin can be determined at any time by monitoring the PROG pin voltage using the following equation:

$$I_{BAT} = \frac{V_{PROG}}{R_{PROG}} \cdot 1000$$

#### **Charge Termination**

A charge cycle is terminated when the charge current falls to 1/10th the programmed value after the final float voltage is reached. This condition is detected by using an internal, filtered comparator to monitor the PROG pin. When the PROG pin voltage falls below 100mV1 for longer than t<sub>TERM</sub> (typically 1ms), charging is terminated. The charge current is latched off and the RY4054 enters standby mode, where the input supply current drops to 200µA. (Note: C/10 termination is disabled in trickle charging and thermal limiting modes).

When charging, transient loads on the BAT pin can cause the PROG pin to fall below 100mV for short periods of time before the DC charge current has dropped to 1/10th the programmed value. The 1ms filter time (t<sub>TERM</sub>) on the termination comparator ensures that transient loads of this nature do not result in premature charge cycle termination. Once the average charge current drops below 1/10th the programmed value, the RY4054 terminates the charge cycle and ceases to provide any current through the BAT pin. In this state, all loads on the BAT pin must be supplied by the battery.

The RY4054 constantly monitors the BAT pin voltage in standby mode. If this voltage drops below the 4.05V recharge threshold ( $V_{RECHRG}$ ), another charge cycle begins and current is once again supplied to the battery. To manually restart a charge cycle when in standby mode, the input voltage must be removed and reapplied, or the charger must be shut down and restarted using the PROG pin. Figure 4 shows the state diagram of a typical charge cycle.

#### Charge Status Indicator (CHRG)

The charge status output has three different states: strong pull-down (~10mA), weak pull-down (~20 $\mu$ A) and high impedance. The strong pull-down state indicates that the RY4054 is in a charge cycle. Once the charge cycle has terminated, the pin state is determined by undervoltage lockout conditions. A weak pull-down indicates that VCC



meets the UVLO conditions and the RY4054 is ready to charge. High impedance indicates that the RY4054 is in undervoltage lockout mode: either VCC is less than 100mV above the BAT pin voltage or insufficient voltage is applied to the VCC pin. A microprocessor can be used to distinguish between these three states—this method is discussed in the Applications Information section.

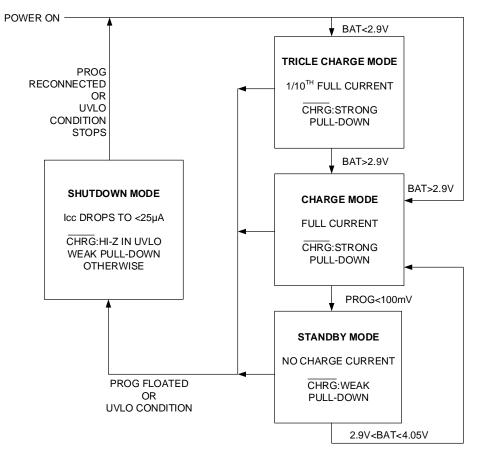


Figure 4. State Diagram of a Typical Charge Cycle

#### Thermal Limiting

An internal thermal feedback loop reduces the programmed charge current if the die temperature attempts to rise above a preset value of approximately 120°C. This feature protects the RY4054 from excessive temperature and allows the user to push the limits of the power handling capability of a given circuit board without risk of damaging the RY4054. The charge current can be set according to typical (not worst-case) ambient temperature with the assurance that the charger will automatically reduce the current in worst-case conditions. ThinSOT power considerations are discussed further in the Applications Information section.

#### **Undervoltage Lockout (UVLO)**

An internal undervoltage lockout circuit monitors the input voltage and keeps the charger in shutdown mode until VCC rises above the undervoltage lockout threshold. The UVLO circuit has a built-in hysteresis of 200mV. Furthermore, to protect against reverse current in the power MOSFET, the UVLO circuit keeps the charger in shutdown mode if VCC falls to within 30mV of the battery voltage. If the UVLO comparator is tripped, the charger will not come out of shutdown mode until VCC rises 100mV above the battery voltage.

#### **Manual Shutdown**



At any point in the charge cycle, the RY4054 can be put into shutdown mode by removing  $R_{PROG}$  thus floating the PROG pin. This reduces the battery drain current to less than  $2\mu$ A and the supply current to less than  $50\mu$ A. A new charge cycle can be initiated by reconnecting the program resistor. In manual shutdown, the *CHRG* pin is in a weak pull-down state as long as VCC is high enough to exceed the UVLO conditions. The *CHRG* pin is in a high impedance state if the RY4054 is in undervoltage lockout mode: either VCC is within 100mV of the BAT pin voltage or insufficient voltage is applied to the VCC pin.

#### Automatic Recharge

Once the charge cycle is terminated, the RY4054 continuously monitors the voltage on the BAT pin using a comparator with a 2ms filter time ( $t_{RECHARGE}$ ). A charge cycle restarts when the battery voltage falls below 4.05V (which corresponds to approximately 80% to 90% battery capacity). This ensures that the battery is kept at or near a fully charged condition and eliminates the need for periodic charge cycle initiations. *CHRG* output enters a strong pulldown state during recharge cycles.

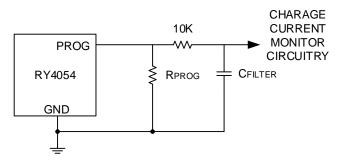
### **Typical Application**

#### **Stability Considerations**

The constant-voltage mode feedback loop is stable without an output capacitor provided a battery is connected to the charger output. With no battery present, an output capacitor is recommended to reduce ripple voltage. When using high value, low ESR ceramic capacitors, it is recommended to add a 1 $\Omega$  resistor in series with the capacitor. No series resistor is needed if tantalum capacitors are used. In constant-current mode, the PROG pin is in the feedback loop, not the battery. The constant-current mode stability is affected by the impedance at the PROG pin. With no additional capacitance on the PROG pin, the charger is stable with program resistor values as high as 20k. However, additional capacitance on this node reduces the maximum allowed program resistor. The pole frequency at the PROG pin should be kept above 100kHz. Therefore, if the PROG pin is loaded with a capacitance, C<sub>PROG</sub>, the following equation can be used to calculate the maximum resistance value for R<sub>PROG</sub>:

$$R_{PROG} \le \frac{1}{2\pi \cdot 10^5 \cdot C_{PROG}}$$

Average, rather than instantaneous, charge current may be of interest to the user. For example, if a switching power supply operating in low current mode is connected in parallel with the battery, the average current being pulled out of the BAT pin is typically of more interest than the instantaneous current pulses. In such a case, a simple RC filter can be used on the PROG pin to measure the average battery current as shown in Figure 5. A 10k resistor has been added between the PROG pin and the filter capacitor to ensure stability.



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Figure 5. Isolating Capacitive Load on PROG Pin and Filtering

#### **Power Dissipation**

The conditions that cause the RY4054 to reduce charge current through thermal feedback can be approximated by considering the power dissipated in the IC. Nearly all this power dissipation is generated by the internal MOSFET—this is calculated to be approximately:

$$P_D = (V_{CC} - V_{BAT}) \cdot I_{BAT}$$

where  $P_D$  is the power dissipated, VCC is the input supply voltage, VBAT is the battery voltage and IBAT is the charge current. The approximate ambient temperature at which the thermal feedback begins to protect the IC is:

$$T_A = 120^{\circ}\text{C} - P_D\theta_{JA}$$
$$T_A = 120^{\circ}\text{C} - (V_{CC} - V_{BAT}) \cdot I_{BAT} \cdot \theta_{JA}$$

Moreover, when thermal feedback reduces the charge current, the voltage at the PROG pin is also reduced proportionally as discussed in the Operation section. It is important to remember that RY4054 applications do not need to be designed for worst-case thermal conditions since the IC will automatically reduce power dissipation when the junction temperature reaches approximately 120°C.

#### **Charge Current Soft-Start**

The RY4054 includes a soft-start circuit to minimize the inrush current at the start of a charge cycle. When a charge cycle is initiated, the charge current ramps from zero to the full-scale current over a period of approximately 100µs. This has the effect of minimizing the transient current load on the power supply during start-up.

#### **CHRG** Status Output Pin

The *CHRG* pin can provide an indication that the input voltage is greater than the undervoltage lockout threshold level. A weak pull-down current of approximately 20µA indicates that sufficient voltage is applied to VCC to begin charging. When a discharged battery is connected to the charger, the constant current portion of the charge cycle begins and the  $\overline{CHRG}$  pin pulls to ground. The  $\overline{CHRG}$  pin can sink up to 10mA to drive an LED that indicates that a charge cycle is in progress. When the battery is nearing full charge, the charger enters the constant-voltage portion of the charge cycle and the charge current begins to drop. When the charge current drops below 1/10 of the programmed current, the charge cycle ends and the strong pull-down is replaced by the 20µA pull-down, indicating that the charge cycle has ended. If the input voltage is removed or drops below the undervoltage lockout threshold, the  $\overline{CHRG}$  pin becomes high impedance. Figure 6 shows that by using two different value pull-up resistors, a microprocessor can detect all three states from this pin. To detect when the RY4054 is in charge mode, force the digital output pin (OUT) high and measure the voltage at the  $\overline{CHRG}$  pin. The N-channel MOSFET will pull the pin voltage low even with the 2k pull-up resistor. Once the charge cycle terminates, the N-channel MOSFET is turned off and a 20 $\mu$ A current source is connected to the  $\overline{CHRG}$  pin. The IN pin will then be pulled high by the 2k pull-up resistor. To determine if there is a weak pull-down current, the OUT pin should be forced to a high impedance state. The weak current source will pull the IN pin low through the 800k resistor; if  $\overline{CHRG}$  is high impedance, the IN pin will be pulled high, indicating that the part is in a UVLO state.



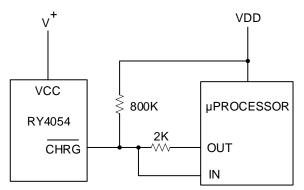


Figure 6. Using a Microprocessor to Determine CHRG State

#### **Reverse Polarity Input Voltage Protection**

In some applications, protection from reverse polarity voltage on VCC is desired. If the supply voltage is high enough, a series blocking diode can be used. In other cases, where the voltage drop must be kept low a P-channel MOSFET can be used (as shown in Figure 7).

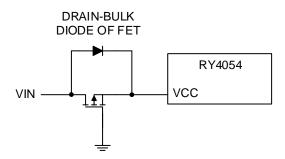


Figure 7. Low Loss Input Reverse Polarity Protection

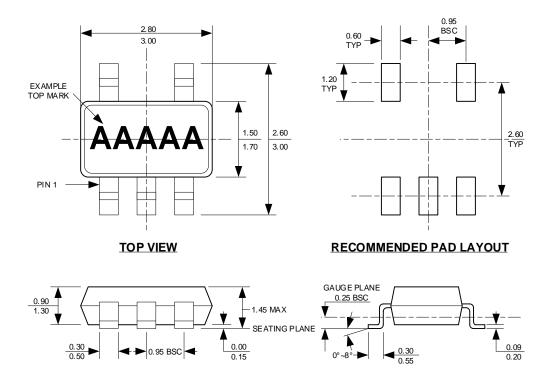
#### **USB and Wall Adapter Power**

The RY4054 allows charging from both a wall adapter and a USB port. Figure 7 shows an example of how to combine wall adapter and USB power inputs. A P-channel MOSFET, MP1, is used to prevent back conducting into the USB port when a wall adapter is present and a Schottky diode, D1, is used to prevent USB power loss through the 1k pull-down resistor. Typically a wall adapter can supply more current than the 500mA-limited USB port. Therefore, an N-channel MOSFET, MN1, and an extra 10k program resistor are used to increase the charge current to 600mA when the wall adapter is present.



### **Package Description**

SOT23-5



FRONT VIEW

SIDE VIEW

NOTE:

1. CONTROL DIMENSION IS IN INCHES. DIMENSION IN BRACKET IS IN MILLIMETERS.

2. PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. 3. PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS. 4. LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.004" INCHES MAX.

5. DRAWING CONFORMS TO JEDEC MS-012, VARIATION BA. 6. DRAWING IS NOT TO SCALE.