

FEATURES

- Low Cost
- Rail-to-Rail Input and Output
 0.8mV Typical V_{OS}
- Unity Gain Stable
- Gain-Bandwidth Product: 100kHz
- Very Low Input Bias Current: 10pA
- Supply Voltage Range: 1.8V to 5.5V
- Input Voltage Range: -0.1V to +5.6V with (V_{DD} - V_{SS}) = 5.5V
- Low Supply Current: 5µA/Amplifier
 - Small Packaging MD1511 Available in SOT-23-5 MD1512 Available in SOP8 MD1514 Available in SOP14

APPLICATIONS

• Current Sensing

- Threshold Detectors/Discriminators
- Low Power Filters
- Handsets and Mobile Accessories
- Wireless Remote Sensors, Active RFID Readers
- Gas/Oxygen/Environment Sensors
- Battery or Solar Powered Devices
- Sensor Network Powered by Energy Scavenging

MD1511/2/4 100kHz, 5µA, Rail-to-Rail I/O CMOS Operational Amplifiers

PRODUCT DESCRIPTION

The MD1511 (single), MD1512 (dual) and MD1514 (quad) are low cost, rail-to-rail input and output voltage feedback amplifiers. They have a wide input common mode voltage range and output voltage swing, and take the minimum operating supply voltage down to 1.8V. The maximum recommended supply voltage is 5.5V. It is specified over the extended -40°C to +85°C temperature range.

The MD1511/2/4 provides 100kHz bandwidth at a low current consumption of 5μ A per amplifier. Very low input bias currents of 10pA enable MD1511/2/4 to be used for integrators, photodiode amplifiers, and piezoelectric sensors. Rail-to-rail input and output are useful to designers for buffering ASIC in single-supply systems.

Applications for this series of amplifiers include safety monitoring, portable equipment, battery and power supply control, and signal conditioning and interfacing for transducers in very low power systems.

The MD1511 is available in the Green SOT-23-5 Package. The MD1512 comes in the Green SOP8 package. The MD1514 comes in the Green SOP14 package.

PIN CONFIGURATIONS (TOP VIEW)









ORDER INFORMATION

MODEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
MD1511		SOT23-5	Tape and Reel, 3000	
MD1512		SOP8	Tape and Reel, 4000	
MD1514		SOP14	Tape and Reel, 4000	

ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V _{DD} to V _{SS}	6V
Common Mode Input Voltage	$\dots V_{SS}$ - 0.3V to V_{DD} + 0.3V
Storage Temperature Range	65℃ to +150℃
Junction Temperature	150°C
Operating Temperature Range	40°C to +85°C

Package Thermal Resistance @ $T_A = +25 ^{\circ}C$
SOP8, θJA125 °C/W
Lead Temperature (Soldering 10sec)260 °C

NOTE:

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ESD, Electrostatic Discharge Protection

Symbol	Parameter	Condition	Minimum Level	Unit
HBM	Human Body Model ESD		1	kV



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ELECTRICAL CHARACTERISTICS

The • denotes the specifications which apply over the full operating temperature range, otherwise specifications are At $T_A=25$ °C, $V_{DD}=+5V$, $V_{SS}=GND$, $R_L = 100k\Omega$, $C_L = 100pF$ connected to $V_{DD}/2$, and $V_{OUT} = V_{DD}/2$.

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	MAX	UNITS	
INPUT CHARACTERISTICS								
	T 7	$V_{CM} = V_{DD}/2$			0.8	5	mV	
Input Offset Voltage	Vos	$V_{CM} = V_{DD}/2$	•			6.6	mV	
Input Bias Current	I _B				10		pA	
Input Offset Current	Ios				10		pA	
Input Offset Voltage Drift	$\Delta V_{OS}/\Delta_T$				3		µV/℃	
Input Common Mode Voltage Range	V _{CM}	$V_{DD} = 5.5 V$			-0.1-5.6		V	
Common Mode Rejection Ratio	CMRR	$V_{\rm DD}$ = 5.5V, $V_{\rm CM}$ = -0.1V to 4V			92		dB	
Open-Loop Voltage Gain	A _{OL}	$R_L = 100 k\Omega, V_{OUT} = 2.5 V$			120		dB	
		R_L = 100kΩ, $V_{\rm OUT}$ = +0.2V to +4.8V			120		dB	
OUTPUT CHARACTERIST	ICS							
	V _{OH}	$R_L = 1M\Omega$		4.980	4.995		V	
	Vol	$R_L = 1M\Omega$		25	5		mV	
Output Voltage Swing from Rail	V _{OH}	$R_L = 100 k\Omega$		4.970	4.994		V	
	V _{OL}	$R_L = 100 k\Omega$		35	6		mV	
	ISOURCE	$R_L = 10\Omega$ to $V_{DD}/2$			80		mA	
Output Current	Isink				70			
POWER SUPPLY								
				1.6		5.5	V	
Operating Voltage Range			•	1.8		5.5	V	
Power Supply Rejection Ratio	PSRR	$V_{DD} = +1.8V$ to $+5.5V$, $V_{CM} = +0.5V$			102		dB	
	Iq				5			
Quiescent Current/Amplifier			•	3.5		6	uA	
DYNAMIC PERFORMANC	$E(C_{L} = 100)$	oF)						
Gain-Bandwidth Product	GBP	$R_L = 1M\Omega$			110			
		$R_L = 100k\Omega$			100		kHz	
Phase Margin	PM	$R_L = 100 k\Omega, C_L = 100 pF$			90		0	
Harmonic Distortion	HD2	$f = 100Hz, G = +1, V_{OUT}=2V_{PP}$			80			
	HD3	$f = 1 kHz, G = +1, V_{OUT}=2V_{PP}$			80		dBc	
Full Power Bandwidth	FPBW				1		kHz	
Slew Rate	SR	G = +1, 2V Output Step			23.7		V/ms	
Settling Time to 0.1%	ts	G = +1, 2V Output Step, 0.1%			26		us	
Overload Recovery Time		$V_{IN} \cdot G = V_{DD}$	1		0.1		ms	
NOISE PERFORMANCE								
Voltage Noise Density	en	f = 1 kHz			50		nV/√F	
		f = 10 kHz			50		nV/√F	



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TYPICAL PERFORMANCE CHARACTERISTICS

At $T_A = +25^{\circ}$ C, $V_{DD} = +5V$, $V_{SS} = GND$, and $R_L = 100k\Omega$ connected to $V_{DD}/2$, unless otherwise specified.



Figure 1. Supply Current vs. Supply Voltage



Figure 3. Output Short Circuit Current vs. Supply Voltage



Figure 5. CMRR vs. Temperature



Figure 2. Supply Current vs. Temperature







Figure 6. PSRR vs. Temperature



TYPICAL PERFORMANCE CHARACTERISTICS

At $T_A = +25^{\circ}$ C, $V_{DD} = +5V$, $V_{SS} = GND$, and $R_L = 100k\Omega$ connected to $V_{DD}/2$, unless otherwise specified.



Figure 7. Small-Signal Overshoot vs. Load Capacitance



Figure 9. Output Voltage vs. Output Current



Figure 11. Input Voltage Noise Spectral Density vs. Frequency



Figure 8. Small-Signal Overshoot vs. Load Capacitance



Figure 10. Output Voltage vs. Output Current



Figure 12. Maximum Output Voltage vs. Frequency



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TYPICAL PERFORMANCE CHARACTERISTICS



Figure 13. Positive Overload Recovery Time



Figure 15. Phase Reversal



Figure 17. Small-Signal Step Response



Figure 14. Negative Overload Recovery Time



Figure 51. Large-Signal Step Response



Figure 18. Small-Signal Step Response



PSRR

CMRR

TYPICAL PERFORMANCE CHARACTERISTICS



Figure 21. Channel Separation vs. Frequency

10 1000 100 Frequency (kHz) Figure 20. CMRR and PSRR vs. Frequency 1000 100 Sol (swyo) 10 1 0.1 0.01 0.1 1 10 100 1000 Frequency (kHz)

Figure 22. Zol vs. Frequency



APPLICATION INFORMATION

MD1511/2/4 are CMOS, rail-to-rail input and output voltage feedback amplifiers designed for general purpose applications. **Operating Voltage**

The MD1511/2/4 are specified over a power-supply range of ± 1.8 V to ± 5.5 V (± 0.9 V to ± 2.75 V), Supply voltages higher than 6V (absolute maximum) can permanently damage the amplifier. Parameters that vary over supply voltage or temperature are shown in the typical characteristics section of this datasheet.

Rail-to-Rail Input

The input stage of the amplifiers is a true rail-to-rail architecture, allowing the input common-mode voltage range of the op amp to extend to both positive and negative supply rails. This maximizes the usable voltage range of the amplifier, an important feature for single-supply and low voltage applications. This rail-to-rail input range is achieved with a complementary input stage—an NMOS input differential pair in parallel with a PMOS differential pair. The NMOS pair is active at the upper end of the common-mode voltage range, typically $V_{DD} - 1.2V$ to 100mV above the positive supply, while the PMOS pair is active for inputs from 100mV below the negative supply to approximately $V_{DD} - 1.2V$.

Rail-to-Rail Output

A class AB output stage with common-source transistors is used to achieve rail-to-rail output. The maximum output voltage swing is proportional to the output current, and larger currents will limit how close the output voltage can get to the proximity of the output voltage to the supply rail. This is a characteristic of all rail-to-rail output amplifiers. See the typical performance characteristic Figure 9, Output Voltage Swing vs. Output Current.

Capacitive Loads

The MD1511/2/4 op amps can directly drive large capacitive loads. As the load capacitance increases, the feedback loop's phase margin decreases and the closed-loop's bandwidth is reduced. This produces gain peaking in the frequency response, with overshoot and ringing in the step response. While a op amp in unity gain configuration (G = +1 V/V) is most susceptible to the effects of capacitive loading.

When driving large capacitive loads with the MD1511/2/4 amplifiers (e.g., > 100pF when G = +1 V/V), a small series resistor at the output (R_{ISO} in Figure 21) improves the feedback



Figure 21. Driving Large Capacitive Loads loop's phase margin (stability) by making the output load resistive at higher frequencies.

PCB Surface Leakage

In Applications where low input bias current is critical, PC board surface leakage effects need to be considered. Surface leakage is caused by humidity, dust or other contamination on the board. Under low humidity conditions, a typical resistance between nearby traces is $10^{12} \Omega$. A 5V difference would cause 5pA of current to flow; which is similar to the MD1511/2/4 op amps' bias current at +25°C (±10pA, typical).

The best way to reduce surface leakage is to use a guard ring around sensitive pins (or traces). The guard ring is biased at the same voltage as the sensitive pin. An example of this type of layout is shown in Figure 22.

- 1. Non-inverting Gain and Unity-Gain Buffer:
 - a) Connect the non-inverting pin (V_{IN+}) to the input with a wire that does not touch the PCB surface.
 - b) Connect the guard ring to the inverting input pin (V_{IN-}) . This biases the guard ring to the Common Mode input voltage.
- 2. Inverting Gain and Transimpedance Gain Amplifiers (convert current to voltage, such as photo detectors):
 - a) Connect the guard ring to the non-inverting input pin (V_{IN+}) . This biases the guard ring to the same reference voltage as the op amp (e.g., $V_{DD}/2$ or ground).
 - b) Connect the inverting pin (V_{IN}) to the input with a wire that does not touch the PCB surface.



Figure 22. Example Guard Ring Layout for Inverting Gain



TYPICAL APPLICATION

Differential Amplifier

The circuit shown in Figure 23 performs the difference function. If the resistor ratios are equal to $(R_4 / R_3 = R_2 / R_1)$, then $V_{OUT} = (V_P - V_N) \times R_2 / R_1 + V_{REF}$.



Figure 23. Differential Amplifier

Photodiode Application

The MD1511/2/4 have very high impedance with an input bias current typically around 10 pA. This characteristic allows the MD1511/2/4 op amp to be used in photodiode applications and other applications that require high input impedance. Note that the MD1511/2/4 have significant voltage offset that can be removed by capacitive coupling or software calibration.

Figure 24 illustrates a photodiode or current measurement application. The feedback resistor is limited to $10 \text{ M}\Omega$ to avoid

excessive output offset. In addition, a resistor is not needed on the noninverting input to cancel bias current offset because the bias current-related output offset is not significant when compared to the voltage offset contribution. For best performance, follow the standard high impedance layout techniques, which include the following:

- Shielding the circuit.
- Cleaning the circuit board.
- Putting a trace connected to the noninverting input around the inverting input.
- Using separate analog and digital power supplies.



Figure 24. High Input Impedance Application—Photodiode Amplifier



RECOMMENDED MOUNTING METHOD

Soldering Methods, Recommended Soldering Method for Moisture-Proof Packing and Flux Cleaning are in the following Mounting was evaluated with the following profiles in our company, so there was no problem. However, confirm mounting by the condition of your company beforehand.

1. Soldering Temperature Profile of Reflow

Recommended reflow soldering temperature profile is in the following



2. Soldering Temperature Profile of Flow

Recommended flow soldering temperature profile is in the following.



3. Soldering Temperature Profile of Iron

Recommended iron soldering temperature profile is in the following.

At 1 lead Temperature: Lower than 350°C Time: within 3s

It is not good for IC's reliability to keep IC high temperature for long time within limit of recommended ranges. Please finish soldering as soon as possible within limit of recommended ranges. See the next section, "IC storage Conditions and Duration" for Moisture-Proof Packing and Deaeration Packing.



^{4.} Note

PACKAGE OUTLINE DIMENSIONS



Figure 25 8-Lead Small Outline Package [SOP8] Dimensions shown in millimeters



PACKAGE OUTLINE DIMENSIONS











PACKAGE OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARD MS-012-AB

Figure 27 14-Lead Small Outline Package [SOP14] Dimensions shown in millimeters

